Digital Oscilloscope

Introduction

- □ A graph-displaying device of electrical signal
	- n. X axis: Time (sometime: voltage)
	- $\overline{\mathbb{R}^n}$ Y axis: Voltage
	- T. Z axis: Intensity or brightness;
- □ Information given by digital oscilloscopes
	- $\overline{}$ Time and voltage
	- $\overline{\mathbb{R}^n}$ Frequency and phase
	- DC and AC components m.
	- $\overline{\mathbb{R}^n}$ Spectral analysis
	- Rise and fall time
	- $\mathcal{C}^{\mathcal{A}}$ Mathematical analysis

Introduction

- □ Control panel of an oscilloscope \Box
	- Т. Vertical Section
	- m, Horizontal Section
	- Trigger Section

\Box Basic setting

- **DEDA** Vertical system T.
	- □ attenuation or amplification of signal \Box (volts/div), signal coupling, etc.
- $\overline{\mathbb{R}^n}$ Horizontal system
	- □ The time base (sec/div), \Box
- $\mathcal{L}_{\mathcal{A}}$ Trigger system
	- \Box To stabilize image for a repeating \Box signal and to trigger on a single event

m. Types of oscilloscopes

- \Box **Analogue oscilloscope** on analogue techniques. It takes signals, amplifying them in an analogue - the traditional form of oscilloscope that relies format and displaying them on a cathode ray tube;
- \Box **Digital storage oscilloscope (DSO) -** the conventional form of digital oscilloscope. It uses a raster type screen to displays the waveform image and other elements (text, cursors, etc);
- **Digital phosphor oscilloscope (DPO):** a highly versatile form of oscilloscope that uses a parallel processing architecture and dedicated processor for waveform acquisition that enable it to capture and display short signals (spurious pulses, glitches and transition errors). It can display the signal in three dimensions: time, amplitude and the values distribution over time (all in real time);
- **Sampling oscilloscope** is used for analyzing very high frequency repetitive signals (higher than the sample rate). It builds up a picture of the waveform with the collected samples from several successive waveforms and assembling them correctly. It use so called equivalent-time sampling technique This oscilloscope bandwidth sometimes can be as high as 50 GHz.

Types of oscilloscopes (cont'd)

$\mathcal{L}(\mathcal{A})$ Types of oscilloscopes (cont'd)
DSO versus

 $107/$ 20 0W Auto $r = 0.08$ -5.89 f $100/$ 0.08 20.00/ Auto -5.80 Ŧ ш Ŧ н e. 100% 2014 Sample Rate = 4.00 GSa/s Main/Delayed Menu Mary/Delayed Menu Sample Rate = 4.006 Sa/s Main Delayed Roll XY. - Time Ref Verrier Main Delayed **Rull** XV. Vernier - Time Ref \checkmark Center Center

DSO versus

DPO

Constant intensity for DSO, proportional with appearance frequency for DPO

DSO

Full-scale (FS) input range (V_{pk-pk}) = 0.2 V, 0.4 V, 1 V, 2 V, or 5 V

Example of DSO technical specification

DSO - main functional blocks

- \Box Signal conditioning
- \Box Acquisition block
- \Box Trigger block
- \Box Base time
- \Box Microprocessor
- \Box **Display**
- \Box Power supply
- \Box Front panel
- \Box **Communication** Interfaces (USB, GPIBRS232, etc.)

Dual channel DSO – block scheme

- Signal conditioning (similar with analog scope)
	- □ Signal coupling (AC, DC , GND)
	- \Box Adding or removing continuous component
	- \Box Filtering, amplification
	- \Box Attenuation with compensation

- \Box Operation realized in analog domain
- \Box Generate syncro pulses (SY)
- \Box Has the same adjustments as analog scope: trigger source, trigger level, slope, hold off timer, signal coupling (AC, DC), trigger signal filtering (direct, LPF, HPF, noise reduction)

- F Acquisition block (real-time sampling)
	- \Box \Box Uniform sampling the signal (fixed rate \mathcal{T}_s)
	- \Box Signal quantization - fast ADC (usually 8 bits)
	- \Box □ Each acquisition cycle (AC) identified by SY command
	- \Box Samplers of acquisition cycle store in memory
	- \Box Acquisition time greater than display time (Ta>Td)
	- \Box Acquisition data: pretrigger / postrigger
	- \Box Acquisition type:
		- Normal (sample);
		- Peak detect;
		- T. Average;

- $\mathcal{L}_{\mathcal{A}}$ Acquisition and display mode:
	- \Box Pre-trigger mode;
	- \Box "Normal" mode;
	- \Box Delay mode;

F Sampling mode

- □ Real-time sampling
	- **4** to 10 samples per period of the highest-frequency component; \blacksquare
	- \mathbf{r} allows pre-trigger mode;
	- allows capturing transients;
- \Box Sequential sampling in equivalent time
	- п for periodic signals only;
	- \mathbf{r} in each period one sample only shifted by ∆t ;
	- \blacksquare ■ equivalent sampling frequency $f_{S,EQ}$ = 1/ Δt ;
- \Box Random sampling in equivalent time
	- \Box for periodic signals only;
	- sampling permanently with maximum sampling frequency (several samples during period of highest-frequency component);
	- **Ta** each set of samples delayed by random but known time;
	- faster reconstruction than sequential sampling;

- F Sample acquisition
	- \Box Display divided in pixels (exp: Nr x Nc – 240 x 250);
	- ❏ Acquiring one sample per display column;
	- ❏ □ Sampling rate function of horizontal scale
		- $N_c^{}$
		- \sim number of horizotal divisions *XN*

□ Maximum sampling rate is technological limited ($f_{\rm Smax}$)

□ does not acquire rapid variations in the signal forfull waveform display;

 \Box may cause narrow pulses to be missed;

• Sample points

Sample mode acquires a single sample point in each interval.

- F Sample acquisition
	- \Box For base time adjustment $C_x < C_x$ min, the sampling rate \Box is limited by $f_{S max}$ and there are insufficient samplers number $(N_A < N_C)$
	- \Box Interpolation
		- **Linear interpolation**; \Box

(for relative many samplers)

 $f_{\rm S} >> 2 \cdot f_m$

 Band-limited interpolation (sinc); $f_S > 2 \cdot f_m$

- $\overline{\mathcal{A}}$ Band-limited interpolation (sinc);
- \mathbb{R}^3 Having samplers $x(nT_s) = x(nNT_{px})$, n={...., -2, -1, 0, 1, 2 ...} the oscilloscope must compute $x(mT_{px})$, $m=\{..., -2, -1, 0, 1, 2...\}$
- $\overline{}$ **Assuming x(t) band-limited signal:** $X(\omega)=0, |\omega| > \omega_{_M}$, $\omega_{_M} < \Omega_{_S}/2$
- $\mathcal{C}^{\mathcal{A}}$ Sampled signal spectrum: $(\omega) = \frac{1}{T} \sum X(\omega - n\Omega_s),$ *s* $\sum_{n=-\infty}^{\infty}$ s $n=-\infty$ $X_e(\omega) = \frac{1}{T_s} \sum_{n=-\infty} X(\omega - n\Omega_s), \quad T_s = \frac{2\pi}{\Omega}$ ∞=−∞ π $\omega = \frac{1}{T_s} \sum X(\omega - n\Omega_s), T_s = \frac{1}{\Omega_s}$ 2 $, \quad \mathbf{1}_{S}$ 1
- T. Recovered signal spectrum: $X(\omega) = T_s \cdot X_e(\omega)$ for $\omega \in \left(-\frac{1}{2}\Omega_s, \frac{1}{2}\Omega_s\right)$ $\left(-\frac{1}{2}\Omega_s,\frac{1}{2}\Omega_s\right)$ 2^{2} 2
- Recovered signal in time domain: $x(t) = \sum_{n=-\infty}^{\infty} x(nT_s) \cdot \operatorname{sinc} \omega_t(t nT_s)$, $\omega_t = \frac{1}{2} \Omega_s = \pi f_s$
(ideal LPF filtering with $f_{\text{cusp}} = f_s(2)$ (ideal LPF filtering with $f_{-3dB} = f_S/2$)
- m. Displayed re-sampled signal:

$$
x\left(mT_{px}\right) = \sum_{n=-\infty}^{\infty} x\left(nT_s\right) \cdot \text{sinc}\,\omega_t\left(m-nN\right)T_{px} = \sum_{n=-\infty}^{\infty} x\left(nNT_{px}\right) \cdot \text{sinc}\,\omega_t\left(m-nN\right)T_{px}
$$
\n
$$
\text{for } m=nN \to x\left(mT_{px}\right) = x\left(nT_s\right)
$$

 \mathbb{R}^3 $\frac{1}{2}$ 2f_m> f_S result in aliasing spectrum and signal distortion;

 $\overline{\mathcal{A}}$ Band-limited v.s. linear interpolation $(C_x < C_x$ _{min})

Assuming the same sample rate and displayed points number

- for $f_{\rm s}$ >>2· $f_m \rightarrow$ identically displayed waveforms;
- for $f_S > 2 \cdot f_m$, but few samples per signal period (<10) \rightarrow different displayed
waveforms: more accurate for band-limited interpolation: waveforms: more accurate for band-limited interpolation;

T, Aliasing phenomena – time domain

(linear or band-limited interpolation and 2 f_m > $f_{\rm S}$);

- Displayed signal has lower frequency than real input signal;

Aliasing phenomena – frequency domain

(band-limited interpolation and 2 f_{m} > $f_{\rm S}$) - Higher frequency componentsof signal are displayed on lower frequency;- Correct displayed componentsNyquist frequency are on the frequency $< f_S / 2$; Amplitude (one-half sample rate) Frequency Aliased frequencies Actual frequencies

- F Peak detection
	- \Box the oscilloscope finds the highest and lowest values of the input signal over each sample interval;
	- **□** Sampling at maximum rate;
	- \Box the oscilloscope can acquire and display narrow pulses;
	- \Box the oscilloscope can identifier aliasing situation
	- \Box Noise will appear

to be higher;

• Sample points displayed

Peak Detect mode displays the highest and lowest acquired voltage in each interval.

 \mathbb{R}^n Peak detection (cont'd)

Noise augmentation

Aliasing detection

Sample mode

Peak detect mode

- $\mathcal{L}(\mathcal{A})$ Average
	- □ Fixed windows \Box
	- \Box Mobile window
	- <mark>□ Continuous</mark>
- reduces asynchronous noise level

Comparative signal displayed in time domain

- $\overline{\mathbb{R}^n}$ **Fixed windows average of length M**
	- \Box *M* acquisition cycles
、
	- <mark>_</mark> Needs high memory
	- \Box \Box Introduces high delay (*M⋅k⋅T_S*)

- F **Fixed windows average of length M**
	- □ Displayed signal values have improved SNR \Box
- $\mathsf{x}_{\textit{disp}}[\mathsf{k}]$ displayed signal samples mean value, in pixel k :
- $\mathsf{x}_{i}[k]$ signal sample in cycle i for pixel k (noise impressed): $\; x_{i}[k] \!=\! u_{i}[k]$ *x* $x_i[k] = u_i[k] + n_i[k]$
- $\omega_{i}[k]$ ideal signal sample in cycle i , the same for all cycles \rightarrow U (synchronized image);
- n [k] $\,$ noise sample in cycle i for pixel k (noise power $\, \sigma_{_n}^2 \,$) $k = \frac{1}{M} \sum u_i[k] + \frac{1}{M} \sum n_i[k] = U + \frac{1}{M} \sum n_i[k]$ 1 \mathbf{M} $_{i=1}$ \mathbf{M} $_{i=1}$ $\frac{1}{2} \sum_{k=1}^{M} \left[k + \frac{1}{2} \sum_{k=1}^{M} n_k[k] = U + \frac{1}{2} \sum_{k=1}^{M} n_k[k]$ $M \stackrel{\text{div}{\mathbf{F}}}{\longrightarrow} M \stackrel{\text{div}{\mathbf{F}}}{\longrightarrow} M \stackrel{\text{div}{\mathbf{F}}}{\longrightarrow} M \stackrel{\text{div}{\mathbf{F}}}{\longrightarrow} M \stackrel{\text{div}{\mathbf{F}}}{\longrightarrow}$ $\sum_{i=1}^{K} \mu_{i} [k] = \frac{1}{M} \sum_{i=1}^{K} \mu_{i} [k] + \frac{1}{M} \sum_{i=1}^{K} \mu_{i} [k] = U + \frac{1}{M} \sum_{i=1}^{K} \mu_{i} [k]$ $=\frac{1}{M}\sum_{i=1}u_i[k]+\frac{1}{M}\sum_{i=1}n_i[k]=U+\frac{1}{M}\sum_{i=1}p_i$
vise power (white noise, uncorrelated *ⁿ*
- output noise power (white noise, uncorrelated with input signal)

$$
\sigma_{n_disp}^{2} = E \left\{ \left(\frac{1}{M} \sum_{i=1}^{M} n_{i}[k] \right)^{2} \right\} = \frac{1}{M^{2}} \sum_{i,j=1}^{M} E \left(n_{i}[k] n_{j}[k] \right) \implies \begin{cases} \sigma_{n_disp}^{2} = \frac{1}{M^{2}} \sum_{i=1}^{M} \sigma_{n}^{2} = \frac{1}{M} \sigma_{n}^{2} \\ \frac{SNR_{disp}}{SNR} = \frac{U^{2}}{\sigma_{no}^{2}} \cdot \frac{\sigma_{n}^{2}}{U^{2}} = M \end{cases}
$$

 $[k] = \frac{1}{M} \sum x_i[k]$

M

1

 $=\frac{1}{M}\sum$

 $\frac{d^{lisp} \mathsf{L}^{\infty} \mathsf{J}^{\perp}}{M} \sum_{i=1}^{J} \mathcal{X}_i \mathsf{L}$ $x_{disp}[k]=\frac{1}{M}\sum_{i=1}^{M}x_{i}[k]$

=

1

=

- \mathbb{R}^n **Nobile window average of length M**
	- \Box *M* acquisition cycles
、
	- \Box Needs high memory
	- \Box Introduces low delay (except for the first cycle)

(memorizing new sample $\;\rightarrow\;$ removing the oldest sample)

$$
x_{disp}^{(i)}[k] = \frac{1}{M} \sum_{l=1}^{M-1} x_{i-l}[k] = x_{disp}^{(i-1)}[k] + \frac{x_i[k]}{M} - \frac{x_{i-M}[k]}{M}
$$

- F Continuous average
	- **n** Needs low memory
	- \Box Introduces low delay (has greatest latency for varying signal)

$$
x_{disp}^{(i)}[k] = \frac{i-1}{i} x_{disp}^{(i-1)}[k] + \frac{1}{i} x_i[k] \quad \text{for } i < M
$$

$$
x_{disp}^{(i)}[k] = \frac{M-1}{M} x_{disp}^{(i-1)}[k] + \frac{1}{M} x_i[k] \quad \text{for } i \ge M
$$

$\mathcal{L}_{\mathcal{A}}$ Average effect – frequency domain

Exp: $for M = 256$ cycles

F Time base

 For analog scope: commanded linear varying voltage generator, (saw-tooth waveform), recycling circuit, OX amplifier, etc;

- □ For digital scope: digital counter that generate the memory addresses of displayed samples \rightarrow the pixels index on the display;
Tangles
- □ Types:
	- $\overline{\mathbb{R}^n}$ ■ Main time base – for normal wave display;
	- $\mathcal{L}_{\mathcal{A}}$ Delayed time base – user controlled delay between display and Sincro time. It operates as controlled zoom function in time domain for postrigger acquired data.

- F Microprocessor operation:
	- \Box Command for user and data interfaces
	- \Box Digital signal processing to improve image quality
	- \Box Signal parameters computing (maximum, minimum, mean, RMS values, signal frequency and period, etc)
	- \Box FFT processing (frequency domain display)
	- \Box Cursors positioning and value displaying
	- \Box Math operation with waveforms (Math menu)
	- \Box Improving SNR by digital filtering for low frequency signal

- F Example: FFT display for Tektronix TDS100
	- \Box **EXTED computation** \rightarrow transforms the center 2048 points in time-domain waveform into 1024 FFT point spectrum;
	- ❏ \Box FFT display: from DC (0Hz) to $f_{\rm S}$ /2 (Nyquist frequency);
	- \Box FFT display – 250 horizontal points (frequency bins);
	- \Box Vertical display - signal harmonic components (rms values);
		- Operation FFT Fundamental frequency CH1 component Window Hanning Frequency component FFT Zoom Ш CH 1 10dB 50.0 kHz (1.00 MS/s) Hanning $\overline{2}$ 3 5 4

Pos:250.0 kHZ

MATH

 $Irria'd$

- 1. Frequency at the center graticule line
- 2. Vertical scale in dB per division (0 dB = 1 V_{RMS})
- 3. Horizontal scale in frequency per division
- 4. Sample rate in number of samples per second
- 5. FFT window type

- F Example: FFT display for Tektronix TDS100
	- \Box FFT options:
		- Source (Ch1, Ch2);
		- FFT zoom (x1, x2, x5, x10);
		- $\overline{\mathbb{R}^n}$ Window
			- □ Hanning (better frequency, poor magnitude accuracy than flattop)
			- □ Flattop (better magnitude, poor frequency accuracy than Hanning)
			- \Box Rectangular (for pulse and transient)

Time-domain

Advantage of Digital Scope

- \blacksquare Trend towards digital.
- \blacksquare Easy to use.
- \blacksquare One-shot measurement
- \blacksquare Recoding (offline processing);
- \blacksquare Triggering (can display pre-triggering waveform);
- \blacksquare Data reuse
- \blacksquare Connectivity (connection interfaces with digital devices –PC). Interfaces: USB, GPIB, RS232;

Others - Probes

- \mathbb{R}^n High quality connector
- \blacksquare High impedance (10MΩ)
- $\mathcal{L}_{\mathcal{A}}$ ■ 50Ω for high frequency measurement

Probes types

- $\mathcal{C}^{\mathcal{A}}$ Passive probes
	- □ 10× attenuation \Box
		- Good for low circuit loading
		- **Suitable to high frequency signal** $\mathcal{L}_{\mathcal{A}}$
		- Difficult to measure less than 10mV signals
	- ¹[×] attenuation
		- Good for small signals
		- **Introducing more interference**
- $\mathcal{C}^{\mathcal{A}}$ Active probes
	- □ Signal conditioning for oscilloscope
	- □ Require power source
	- □ Good for high speed digital signals over 100MHz clock frequency

$\overline{\mathbb{R}^n}$ **Bibliography**

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